# 1000-MHz Quadrature Modulator

### **Description**

U2790B is a 1000-MHz quadrature modulator that uses TELEFUNKEN's advanced UHF process. It features a frequency range from 100 MHz up to 1000 MHz, low current consumption, and single-ended RF and LO ports.

Adjustment free application makes the direct converter suitable for all digital radio systems up to 1000 MHz, e.g., GSM, ADC, JDC.

#### **Features**

- Supply voltage 5 V (typical)
- Very low power consumption: 150 mW (typical) for -1 dBm output level
- Very good sideband suppression by means of duty cycle regeneration of the LO input signal
- Phase control loop for precise 90° phase shifting
- Power down mode
- Low LO input level: -10dBm (typical)
- $50-\Omega$  single-ended LO and RF port
- LO- frequency from 100 MHz to 1 GHz
- SO 16 package

### **Benefits**

- No external components required for phase shifting
- Adjustment free, hence saves time
- Only three external components result in cost and board space saving

### **Block Diagram**

U2790B-FP

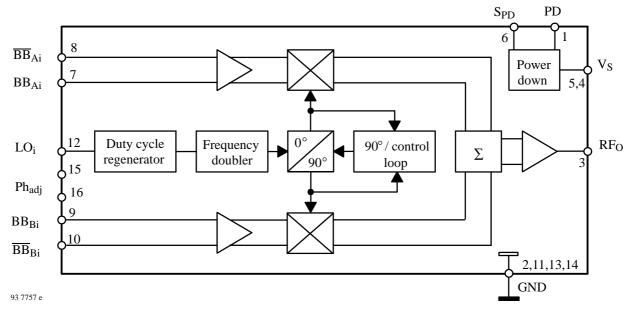


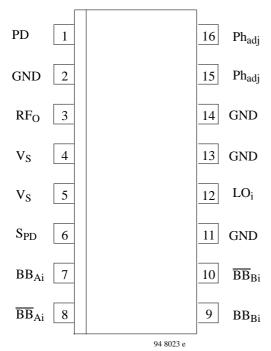
Figure 1.

# **U2790B**

**TELEFUNKEN Semiconductors** 

## **Pin Description**

SO 16



### U2790B-FP (SO 16)

Pin	Symbol	Function	
1	PD	Power down port	
2, 11,	GND	Ground	
13, 14			
3	$RF_{o}$	RF output	
4, 5	V <sub>S</sub>	Supply voltage	
6	$S_{PD}$	Settling time power down	
7	$BB_{Ai}$	Baseband input A	
8	$\overline{BB}_{Ai}$	Baseband input A inverse	
9	$BB_{Bi}$	Baseband input B	
10	$\overline{BB}_{Bi}$	Baseband input B inverse	
12	LOi	LO input	
15/16	Ph <sub>adj</sub>	Phase adjustment (not neces-	
		sary for regular applications)	

# **Absolute Maximum Ratings**

Parameters		Symbol	Value	Unit
Supply voltage	Pins 4 and 5	$V_{S}$	6	V
Input voltage	Pins 7, 8, 9, 10 and 12	V <sub>i</sub>	0 to V <sub>S</sub>	V
Junction temperature		Ti	125	°C
Storage temperature range		T <sub>stg</sub>	-40  to + 125	°C

# **Operating Range**

Parameters	Symbol	Value	Unit
Supply voltage range Pins 4 and 5	$V_{S}$	4.5 to 5.5	V
Ambient temperature range	T <sub>amb</sub>	-40  to  +85	°C

## **Thermal Resistance**

Parameters		Symbol	Value	Unit
Junction ambient	SO 16	$R_{thJA}$	110	K/W

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### **Electrical Characteristics**

Test conditions (unless otherwise specified):  $V_S = 5~V$ ,  $T_{amb} = 25^{\circ}C$ , referred to test circuit, system impedance  $Z_O = 50~\Omega$ ,  $f_{LO} = 900~MHz$ ,  $P_{LO} = -10~dBm$ ,  $V_{BBi} = 1~V_{pp}$  diff

Parameters	Test Conditions / Pin	Symbol	Min.	Тур.	Max.	Unit
Supply voltage range	Pins 4 and 5	V <sub>S</sub>	4.5		5.5	V
Supply current	Pins 4 and 5	Is		30		mA
<b>Baseband inputs</b>	Pins 7-8, 9-10					•
Input voltage range		$V_{\mathrm{BBi}}$		1000	1500	mV <sub>pp</sub>
(differential)						TT
Input impedance		$Z_{BBi}$		3.2		kΩ
(single ended)						
Input frequency range		$f_{BBi}$	0		200	MHz
Internal bias voltage		$V_{\mathrm{BBb}}$	2.35	2.5	2.65	V
Temperature coefficient		$TC_{BB}$		0.1	<1	mV/°C
LO input	Pin 12					
Frequency range		$f_{LOi}$	100		1000	MHz
Input level <sup>1</sup>		P <sub>LOi</sub>	-12	-10	-5	dBm
Input impedance		Z <sub>iLO</sub>		50		Ω
Voltage standing wave ratio		VSWR <sub>LO</sub>		1.4	2	_
Duty cycle range		DCR <sub>LO</sub>	0.4		0.6	-
RF output	Pin 3			•		•
Output level		P <sub>RFo</sub>	-5	-1		dBm
LO suppression <sup>2</sup>	$f_{LO:} = 900 \text{ MHz}$	LO <sub>RFo</sub>	30	35		dB
	$f_{LO:} = 150 \text{ MHz}$		32	35		
Sideband suppression <sup>2,3</sup>	$f_{LO:} = 900 \text{ MHz}$	SBS <sub>RFo</sub>	35	40		dB
	$f_{LO:} = 150 \text{ MHz}$		30	35		
Phase error <sup>4</sup>		Pe		< 1		deg.
Amplitude error		A <sub>e</sub>		$< \pm 0.25$		dB
Noise floor	$V_{BBi} = 2 \text{ V}, \overline{V}_{BBi} = 3 \text{ V}$ $V_{BBi} = \overline{V}_{BBi} = 2.5 \text{ V}$	$N_{\mathrm{FL}}$		- 132 - 144		dBm/Hz
VSWR		VSWR <sub>RF</sub>		1.6	2	
3rd order baseband harmonic suppression		S <sub>BBH</sub>	35	45		dB
RF harmonic suppression		S <sub>RFH</sub>		35		dB
Power down mode	1	11111		l		
Supply current	$V_{PD} \le 0.5 \text{ V}$ Pins 4, 5 $V_{PD} = 1 \text{ V}$	I <sub>PD</sub>		10	1	μΑ
Settling time	$C_{SPD} = 100 \text{ pF}$ $C_{LO} = 100 \text{ pF}$ $C_{RFo} = 1 \text{ nF} \qquad \text{Pin 6 to 3}$	t <sub>sPD</sub>		10		μs
Switching voltage	Pin 1					
Power on		V <sub>PDon</sub>	4			V
Power down		V <sub>PDdown</sub>			1	V

Note: 1 The required LO level is a function of the LO frequency.

Note: 2 In reference to a RF output level  $\leq -1$  dBm and I/Q input level of 400 mV<sub>pp</sub> diff

Note: 3 Sideband suppression is tested without connection at Pins 15 and 16.

For higher requirements a potentiometer can be connected at these pins.

Note: 4 For  $T_{amb} = -30 \text{ to} + 85^{\circ}\text{C}$  and  $V_S = 4.5 \text{ to } 5.5 \text{ V}$ 

Typical Single Sideband Output Spectrum at  $V_S$  =4.5 V and  $V_S$  = 5.5 V  $f_{LO}$  = 900 MHz,  $P_{LO}$  = - 10 dBm,  $V_{BBi}$  = 1  $V_{PP}$  (differential)  $T_{amb}$  = 25°C

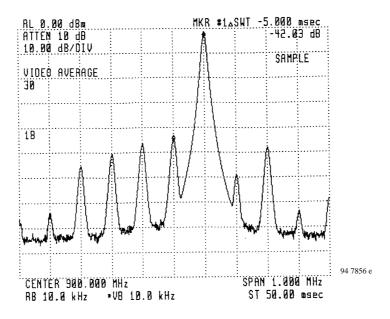


Figure 2.

# **Typical GMSK Output Spectrum**

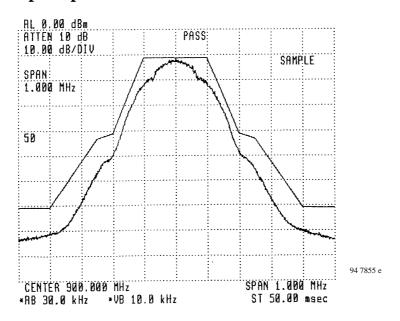


Figure 3.

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## **Typical RF-Harmonic Output Spectrum**

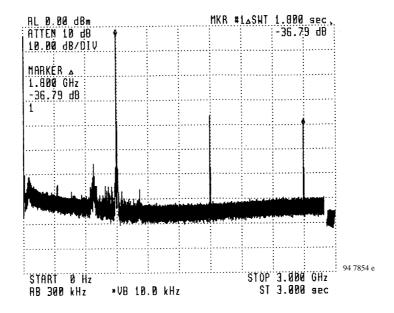
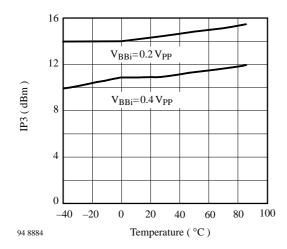


Figure 4.



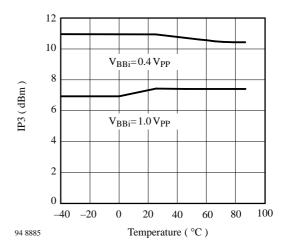
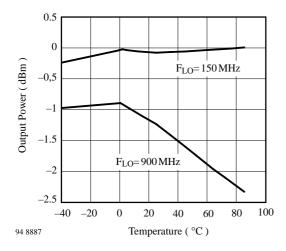


Figure 5. OIP3 vs.  $T_{amb}$ , LO = 150 MHz, level – 20 dBm

Figure 6. OIP3 vs.  $T_{amb}$ , LO = 900 MHz, level – 10 dBm



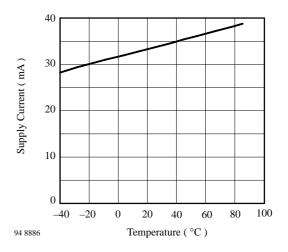


Figure 7. Output power vs. T<sub>amb</sub>

Figure 8. Supply current vs. T<sub>amb</sub>

## Typical S11 Frequency Response of the RF Output

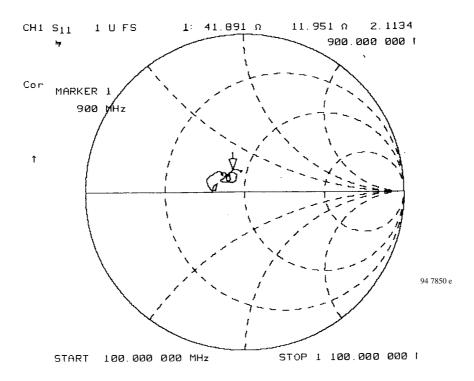
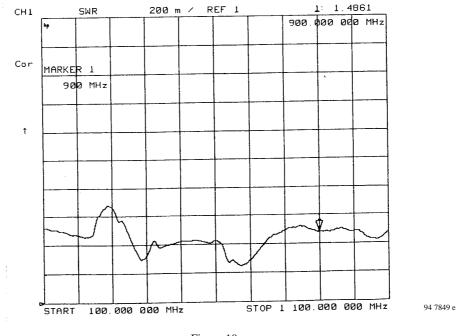


Figure 9.

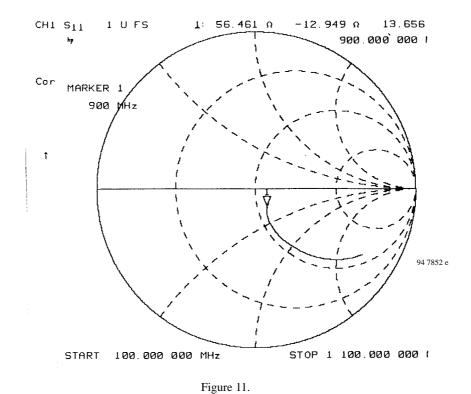
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## Typical VSWR Frequency Response of the RF Output



### Figure 10.

# **Typical S11 Frequency Response of the LO Input**



Rev. A2: 19.01.1995

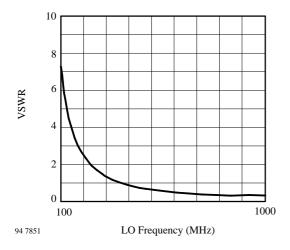


Figure 12. Typical VSWR frequency response of the LO input

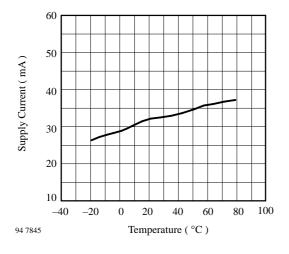


Figure 13. Typical supply current vs. temperature at  $V_S = 5 \text{ V}$ 

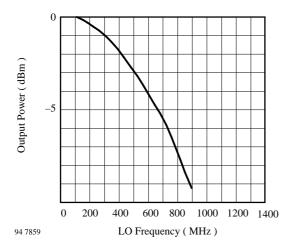


Figure 14. Typical output power vs. LO-frequency at  $T_{amb}$  = 25°C,  $V_{BBi}$  = 230 mV<sub>PP</sub> (differential)

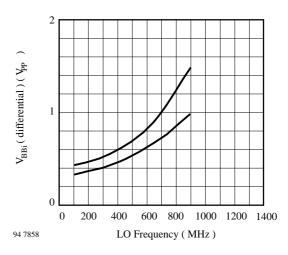


Figure 15. Typical required  $V_{BBi}$  input signal (differential) vs. LO frequency for  $P_O=0\ dBm$  and  $P_O=-2\ dBm$ 

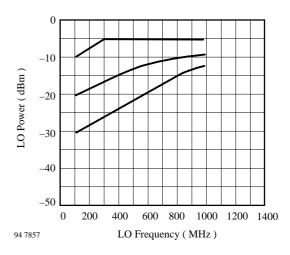


Figure 16. Typical useful LO power range vs. LO frequency at  $$T_{amb}=25\ ^{\circ}C$$ 

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# **Application Circuit**

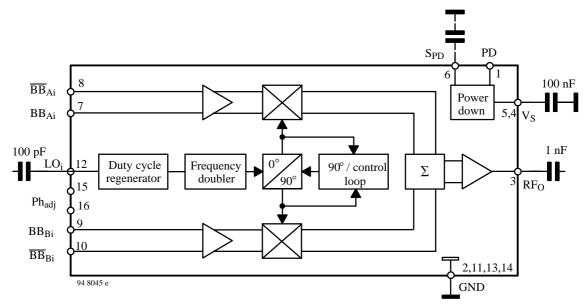


Figure 17.

# **PCB** Layout

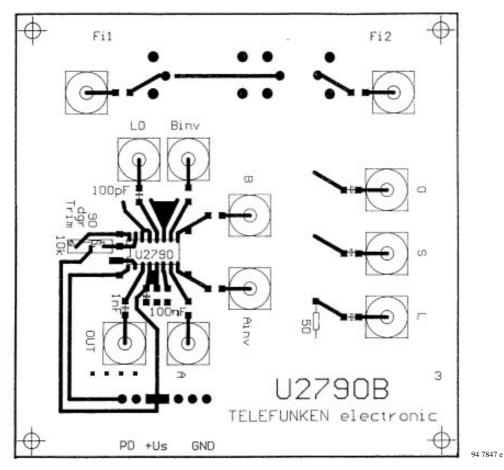


Figure 18. U2790B-FP (SO 16)

### **Application Notes**

#### 1. Noise floor and settling time

In order to reduce noise on the power down control input and improve the wide-off noise floor of the 900-MHz RF output signal, capacitor  $C_{PD}$  should be connected from Pin 6 to ground in the shortest possible way.

The settling time has to be considered for the system under design. For GSM applications a value of  $C_{PD}=1~nF$  defines a settling time,  $t_{sPD}$ , equal or less than 3  $\mu s$ . This capacitance does not have any influence on the noise floor within the relevant GSM mask. For mobile application

the mask requirements can be achieved very easily without C<sub>PD</sub>.

A significant improvement of the wide-off noise floor is obtainable with  $C_{PD}$  greater than 100 nF. Such values are recommended for applications where the settling time is not critical, such as in base stations. Coupling capacitors for  $LO_i$  and  $RF_O$  also have a certain impact on the settling time. The values used for the measurements are  $C_{LOi} = 100 \ pF$  and  $C_{RFo} = 1 \ nF$ 

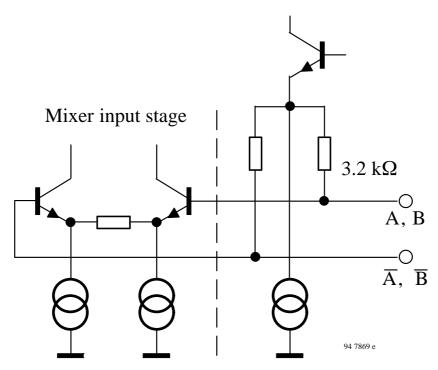


Figure 19. Baseband input circuitry

### 2. Baseband coupling

U2790B-FP (SO 16) has an integrated biasing network which allows ac coupling of the baseband signal at a low count of external components. The bias voltage is  $2.5~V\pm0.15~V$ .

Figure 1 shows the baseband input circuitry with a resistance of 3.2 k $\Omega$  for each asymmetric input. The internal dc offset between A and  $\overline{A}$ , and B and  $\overline{B}$  is typically  $<\pm$  1 mV with a maximum of  $\pm$  3 mV. DC coupling is also possible with an external dc voltage of 2.5  $\pm$  0.15 V.

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### 3. Circuitries

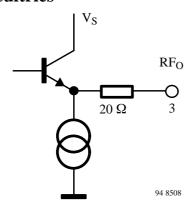


Figure 20. RF output circuitry

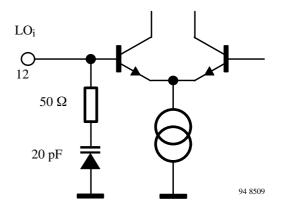


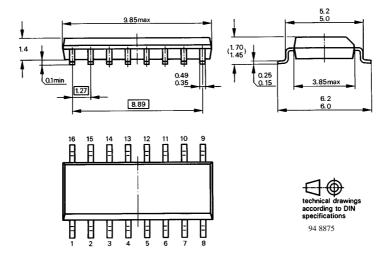
Figure 21. L<sub>O</sub> input circuitry

# **Ordering Information**

Package	Part number
SO 16	U2790B-FP

## **Dimensions in mm**

Package: SO 16



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### **Ozone Depleting Substances Policy Statement**

It is the policy of TEMIC TELEFUNKEN microelectronic GmbH to

- 1. Meet all present and future national and international statutory requirements.
- Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

**TEMIC TELEFUNKEN microelectronic GmbH** semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

- 1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
- 2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
- 3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

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